Adaptive filtering in the complex wavelet domain with unary filters: application to multiple suppression in geophysics On echoes and morphing

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On echoes and morphing

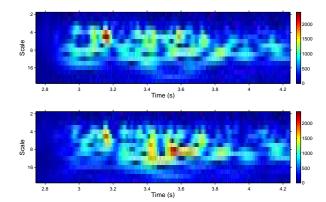


Figure 1: ... and adaptive subtraction



Agenda

- 1. Issues in geophysical signal processing
- 2. Problem: multiple reflections (echoes)
 - adaptive filtering with approximate models
- 3. Complex, continuous wavelets
 - and how they (may) simplify adaptive filtering
- 4. Discretization, redundancy and unary filters (morphing)
 - being practical: back to the discrete world
- Results
- 6. Conclusion



Results & conclusion

Context

Issues in geophysical signal processing

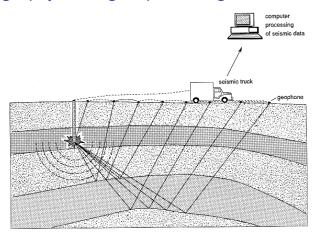


Figure 2: Seismic data acquisition and wave fields.



Issues in geophysical signal processing

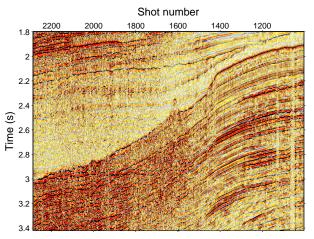


Figure 3: Seismic data: aspect & dimensions (time, offset)



Issues in geophysical signal processing

Reflection seismology:

- seismic waves propagate through the subsurface medium
- seismic traces: seismic wave fields recorded at the surface
 - primary reflections: geological interfaces
 - many types of distortions/disturbances
- processing goal: extract relevant information for seismic data
- led to important signal processing tools:
 - ℓ₁-promoted deconvolution (Claerbout, 1973)
 - wavelets (Morlet, 1975)
- exabytes (10^6 gigabytes) of incoming data
 - need for fast, scalable algorithms



Multiple reflections and models

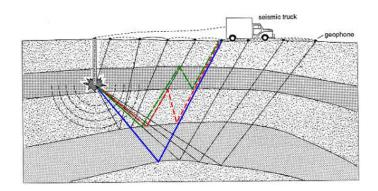


Figure 4: Seismic data acquisition: focus on multiple reflections



Multiple reflections:

- seismic waves bouncing between layers
- one of the most severe types of interferences
- obscure deep reflection layers
- high cross-correlation between primaries (p) and multiples (m)
- additional incoherent noise (n)
- d(t) = p(t) + m(t) + n(t)
 - model-based multiple attenuation: $x_1(t)$, $x_2(t)$, $x_3(t)$
- how to use approximate models?



Multiple reflections and models

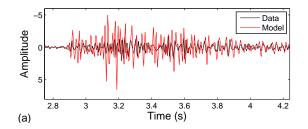


Figure 5: Multiple reflections: data trace d and model x_1



Multiple reflections and models

Multiple filtering:

- multiple prediction (correlation, wave equation) has limitations
- models are not accurate
 - $m(t) \approx a_k(t)x_k(t-\tau_k(t))$?
 - standard: identify, apply a matching filer, subtract
- primaries and multiples are not (fully) uncorrelated
 - same (seismic) source
 - similarities/dissimilarities in time
 - similarities/dissimilarities in frequency
- variations in amplitude, waveform, delay
- issues in matching filter length:
 - short filters and windows: local details
 - long filters and windows: large scale effects



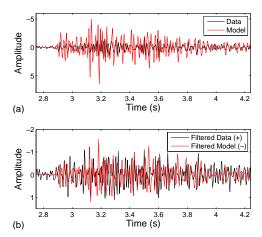


Figure 6: Multiple reflections: data trace, model and adaptation



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Multiple reflections and models

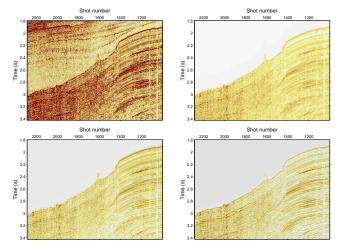


Figure 7: Multiple reflections: data trace and models, 2D version



Multiple reflections and models

- A long history of multiple filtering methods
 - general idea: combine adaptive filtering and transforms
 - data transforms: Fourier. Radon
 - enhance the differences between primaries, multiples and noise
 - reinforce the adaptive filtering capacity
 - intrication with adaptive filtering?
 - might be complicated (think about inverse transform)
- Main idea here:
 - exploit the non-stationary in the data
 - naturally allow both large scale & local detail matching
 - work in a complex domain: amplitude and phase representation
 - emulate an analytic signal representation (Hilbert transform)
- ⇒ Complex, continuous wavelets
 - intermediate complexity in the transform
 - hyper-simplicity in the (unary) adaptive filtering



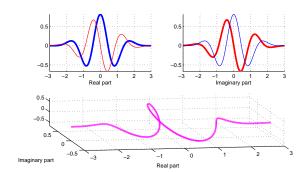


Figure 8: Complex wavelets at two different scales - 1



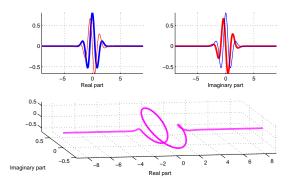


Figure 9: Complex wavelets at two different scales - 2



Transformation group:

affine = translation
$$(\tau)$$
 + dilation (a)

Basis functions:

$$\psi_{\tau,a}(t) = \frac{1}{\sqrt{a}}\psi\left(\frac{t-\tau}{a}\right)$$

- *a* > 1: dilation
- *a* < 1: contraction
- $1/\sqrt{a}$: energy normalization
- multiresolution (vs monoresolution in STFT)

$$\psi_{\tau,a}(t) \xrightarrow{\mathrm{FT}} \sqrt{a}\Psi(af)e^{-i2\pi f\tau}$$



Definition

$$C_s(\tau, a) = \int s(t)\psi_{\tau, a}^*(t)dt$$

Discret., redundancy, unary filters

Vector interpretation

$$C_s(\tau, a) = \langle s(t), \psi_{\tau, a}(t) \rangle$$

projection onto time-scale atoms (vs time-frequency)

- Redundant transform: $\tau \to \tau \times a$ "samples"
- Parseval-like formula

$$C_s(\tau, a) = \langle X(f), \Psi_{\tau, a}(f) \rangle$$

 \Rightarrow time-scale domain operations! (cf. Fourier)



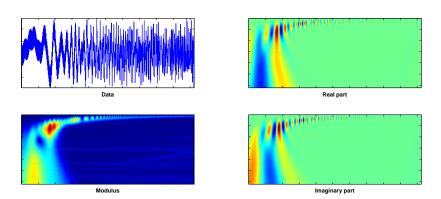


Figure 10: Noisy chirp mixture in time-scale & sampling



Noise spread & feature simplification

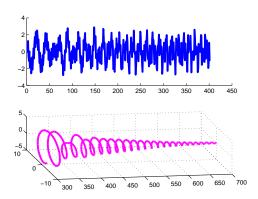


Figure 11: Noisy chirp mixture in time-scale: scale, zoomed wiggle



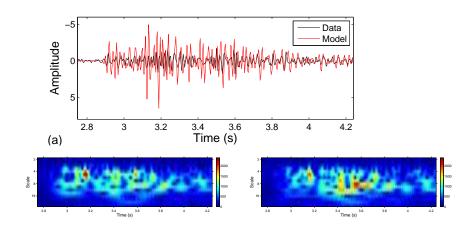


Figure 12: Which morphing is easier: time or time-scale?



• Inversion with another wavelet ϕ

$$s(t) = \iint C_s(u, a)\phi_{u, a}(t) \frac{duda}{a^2}$$

Discret., redundancy, unary filters

⇒ time-scale domain processing! (back to the signal)

Scalogram

$$|C_s(t,a)|^2$$

Energy conversation

$$E = \iint |C_s(t,a)|^2 \frac{dtda}{a^2}$$

Parseval-like formula

$$\langle s_1, s_2 \rangle = \iint C_{s_1}(t, a) C_{s_2}^*(t, a) \frac{dtda}{a^2}$$



Wavelet existence: admissibility criterion

$$0 < A_h = \int_0^{+\infty} \frac{\widehat{\Phi}^*(\nu)\Psi(C)}{\nu} d\nu = \int_{-\infty}^0 \frac{\widehat{\Phi}^*(\nu)\Psi(\nu)}{\nu} d\nu < \infty$$

Discret., redundancy, unary filters

generally normalized to 1

- Induces band-pass property:
 - necessary condition: $|\Phi(0)| = 0$, or zero-average shape
 - amplitude spectrum neglectable w.r.t. |v| at infinity
- examples: Morlet-Gabor (non. adm.)

$$\psi(t) = \frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{t^2}{2\sigma^2}} e^{-i2\pi f_0 t}$$



Discretization, redundancy and unary filters

Being practical again: deal with discrete signals

• Can one sample in time-scale (CWT):

$$C_s(\tau, a) = \int s(t)\psi_{\tau, a}^*(t)dt, \quad \psi_{\tau, a}(t) = \frac{1}{\sqrt{a}}\psi\left(\frac{t - \tau}{a}\right)$$

with $c_{j,k} = C_s(kb_0a_0^j, a_0^j), \ (j,k) \in \mathbb{Z}$ and still be able to recover s(t)?

- Result 1 (Daubechies, 1984): there exists a wavelet frame if $a_0b_0 < C$, (depending on ψ). A frame is generally redundant
- Result 2 (Meyer, 1985): there exist an orthonormal basis for a specific ψ (non trivial, Meyer wavelet) and $a_0=2$ $b_0=1$

Now: how to choose the practical level of redundancy?



Discretization, redundancy and unary filters

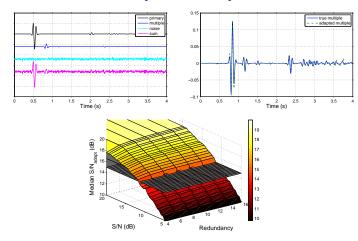


Figure 13: Redundancy selection with variable noise experiments



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Discretization, redundancy and unary filters

Complex Morlet wavelet:

$$\psi(t) = \pi^{-1/4} e^{-i\omega_0 t} e^{-t^2/2}, \, \omega_0$$
: central frequency

Discretized time r, octave j, voice v:

$$\psi_{r,j}^v[n] = \frac{1}{\sqrt{2^{j+v/V}}} \psi\left(\frac{nT - r2^j b_0}{2^{j+v/V}}\right), b_0$$
: sampling at scale zero

Time-scale analysis:

$$\mathbf{d} = d_{r,j}^v = \left\langle d[n], \psi_{r,j}^v[n] \right\rangle = \sum_n d[n] \overline{\psi_{r,j}^v[n]}$$



Discretization, redundancy and unary filters

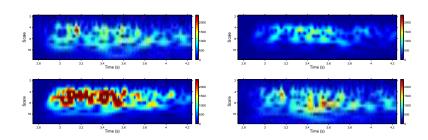


Figure 14: Morlet wavelet scalograms, data and models

Take advantage from the closest similarity/dissimilarity:

• remember the wiggle: on sliding windows, at each scale, a single complex coefficient compensates amplitude and phase

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Discretization, redundancy and unary filters

 Windowed adaptation: complex a_{opt} compensates local delay/amplitude mismatches:

$$\mathbf{a}_{\text{opt}} = \underset{\{a_k\}(k \in K)}{\operatorname{arg min}} \left\| \mathbf{d} - \sum_{k} a_k \mathbf{x}_k \right\|^2$$

Vector Wiener equations for complex signals:

$$\langle \mathbf{d}, \mathbf{x}_m \rangle = \sum_k a_k \langle \mathbf{x}_k, \mathbf{x}_m \rangle$$

Time-scale synthesis:

$$\hat{d}[n] = \sum_{r} \sum_{i,v} \frac{\hat{d}^{v}_{r,j}}{\tilde{\psi}^{v}_{r,j}}[n]$$



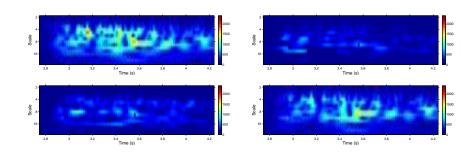


Figure 15: Wavelet scalograms, data and models, after unary adaptation



Results (reminders)

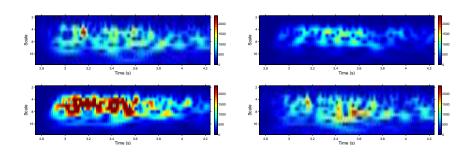


Figure 16: Wavelet scalograms, data and models



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Results

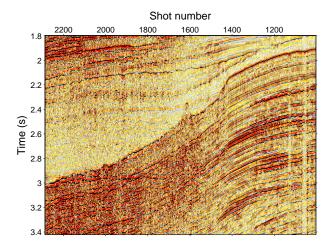


Figure 17: Original data



Results

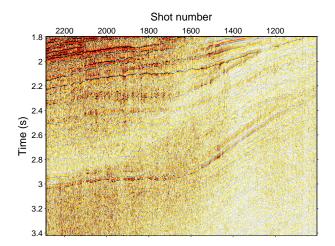


Figure 18: Filtered data



Conclusions

Take-away messages:

- Technical side
 - Take good care of cascaded processing
 - Non-stationary, wavelet-based, adaptive multiple filtering
 - "Complex" wavelet transform + simple one-tap (unary) filter
 - Redundancy selection: noise robustness and processing speed
 - Smooth adaptation to adaptive joint multiple model filtering
- Practical side
 - Industrial integration
 - Competitive with more standard processing
 - Alternative results: less sensitive to random noises
- Future work: better integrate incoherent noise



Acknowledgements & references







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